

8-bit 20MSPS Video A/D Converter with Clamp Function

Description

The CXD1176Q is an 8-bit CMOS A/D converter for video use that features a sync clamp function. The adoption of a 2 step-parallel method realizes low power consumption and a maximum conversion speed of 20MSPS.

Features

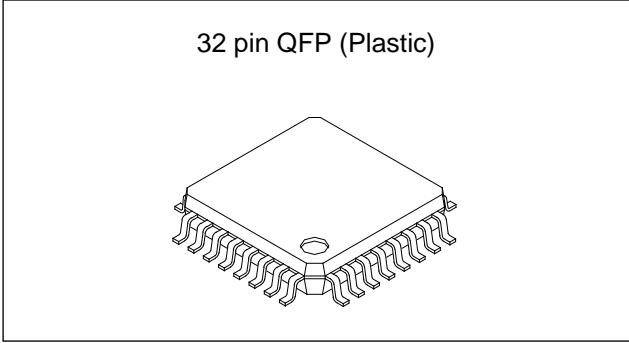
- Resolution power: 8-bit $\pm 1/2$ LSB (DL)
- Maximum sampling frequency: 20MSPS
- Low power consumption: 60mW (at 20MSPS typ.) (Reference current excluded)
- Built-in sync type clamp function
- Built-in monostable multivibrator for clamp pulse generation
- Built-in sync pulse polarity selection function
- Clamp pulse direct input possible
- Built-in clamp ON/OFF function
- Built-in reference voltage self-bias circuit
- Input CMOS compatible
- 3-state TTL compatible output
- Single 5V power supply
- Low input capacity: 11pF
- Reference impedance: 330 Ω (typ.)

Applications

TV and VCR digital systems and a wide range of applications where high-speed A/D conversion is required.

Structure

Silicon gate CMOS IC



Absolute Maximum Ratings (Ta = 25°C)

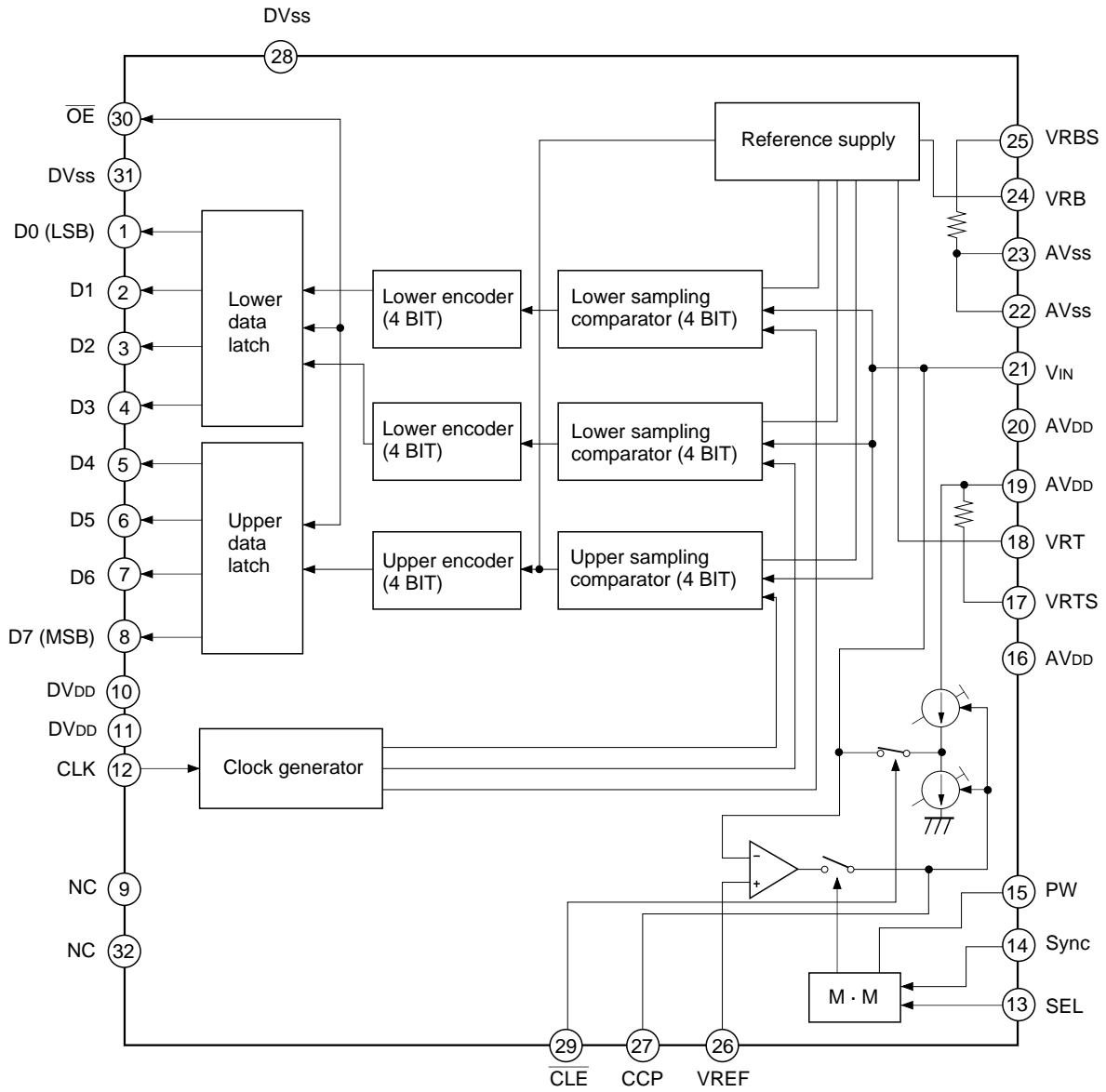
- Supply voltage V_{DD} 7 V
- Reference voltage V_{RT}, V_{RB} $V_{DD} + 0.5$ to $V_{SS} - 0.5$ V
- Input voltage V_{IN} $V_{DD} + 0.5$ to $V_{SS} - 0.5$ V (Analog)
- Input voltage V_I $V_{DD} + 0.5$ to $V_{SS} - 0.5$ V (Digital)
- Output voltage V_O $V_{DD} + 0.5$ to $V_{SS} - 0.5$ V (Digital)
- Storage temperature T_{stg} -55 to +150 °C

Recommended Operating Conditions

- Supply voltage AV_{DD}, AV_{SS} 4.75 to 5.25 V
 DV_{DD}, DV_{SS} $|DV_{SS} - AV_{SS}|$ 0 to 100 mV
- Reference input voltage V_{RB} 0 to V
 V_{RT} to 2.7 V
- Analog input V_{IN} 1.8Vp-p above
- Clock pulse width T_{pw1}, T_{pw0} 22.5ns (min) to 1.1 μ s (max)
- Operating ambient temperature T_{opr} -40 to +85 °C

Sony reserves the right to change products and specifications without prior notice. This information does not convey any license by any implication or otherwise under any patents or other right. Application circuits shown, if any, are typical examples illustrating the operation of the devices. Sony cannot assume responsibility for any problems arising out of the use of these circuits.

Block Diagram and Pin Configuration



Pin Description

Pin No.	Symbol	Equivalent circuit	Description
1 to 8	D0 to D7		D0 (LSB) to D7 (MSB) output
9, 32	NC		NC pin
10, 11	DV _{DD}		Digital +5V
12	CLK		Clock input
13	SEL		<p>When SEL is at low, with the falling edge of Pin 14 (sync) as trigger, the monostable multivibrator generates clamp pulses.</p> <p>When SEL is at high, with the rising edge of Pin 14 (sync) as trigger, it generates clamp pulses.</p>
14	Sync		<p>Trigger pulse input to the monostable multivibrator.</p> <p>Trigger polarity can be selected through Pin 13 (SEL).</p>

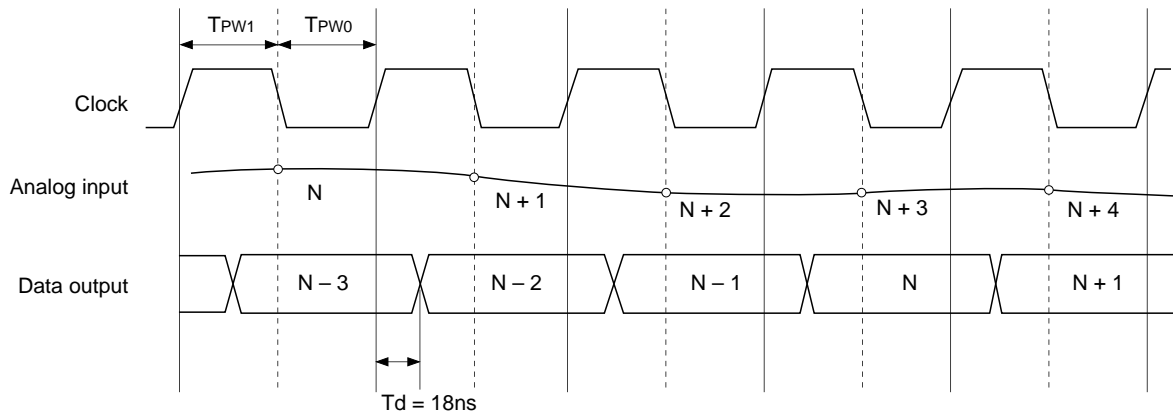
Pin No.	Symbol	Equivalent circuit	Description
15	PW		<p>When a clamp pulse is generated at the monostable multivibrator, the pulse width is determined by the external R and C.</p> <p>When the clamp pulse is directly input, it is input to Pin 15 (PW). The signal voltage of the low period is clamped. (Here, Pin 14 (sync) is fixed to either low or high.)</p>
16, 19, 20	AVDD		Analog +5V
17	VRTS		When shorted with VRT, generates approx. +2.6V.
18	VRT		Reference voltage (top)
24	VRB		Reference voltage (bottom)
21	V _{IN}		Analog input
22, 23	AVSS		Analog ground
25	VRBS		When shorted with VRB, generates approx. +0.5V.

Pin No.	Symbol	Equivalent circuit	Description
26	VREF		Clamp reference voltage input. Clamps to provide a clamp period input signal equal to the reference voltage.
27	CCP		Integrates the voltage for clamp control. CCP and VIN voltage changes are in positive phase.
28, 31	DVss		Digital ground.
29	$\overline{\text{CLE}}$		When $\overline{\text{CLE}}$ is at low, clamp function is activated. When $\overline{\text{CLE}}$ is at high, clamp function is OFF and only the usual A/D converter function is active. By connecting $\overline{\text{CLE}}$ pin to DVDD via a several hundred Ω resistance, the clamp pulse can be tested.
30	$\overline{\text{OE}}$		When $\overline{\text{OE}}$ is at low, Data is output. When $\overline{\text{OE}}$ is at high, D0 to D7 pins turn to high impedance.

Digital Output

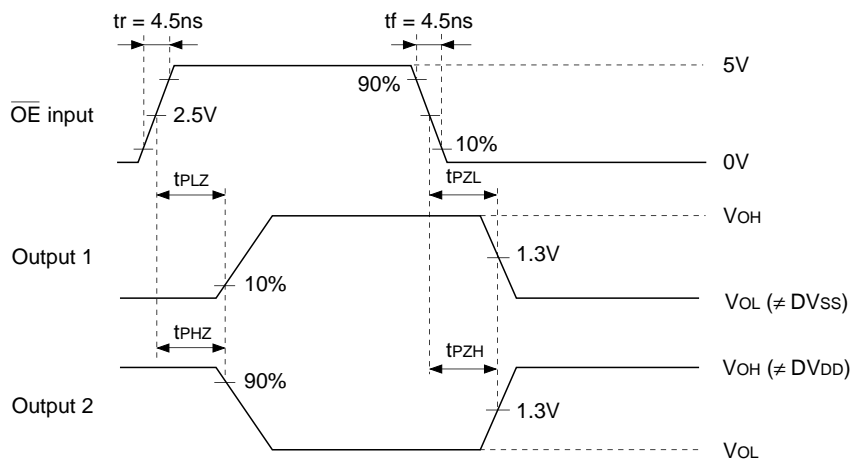
Correspondence between the analog input voltage and the digital output code is indicated in the chart below.

Input signal voltage	Step	Digital output code	
		MSB	LSB
V_{RT}	0	1 1 1 1 1 1 1 1	
⋮	⋮	⋮	
	127	1 0 0 0 0 0 0 0	
⋮	⋮	⋮	
	128	0 1 1 1 1 1 1 1	
⋮	⋮	⋮	
V_{RB}	255	0 0 0 0 0 0 0 0	



○ : Points where analog signals are sampled.

Timing Chart. I



Timing Chart. II

Electrical Characteristics

Analog characteristics

(F_c = 20MSPS, V_{DD} = 5V, V_{RB} = 0.5V, V_{RT} = 2.5V, T_a = 25°C)

Item	Symbol	Conditions	Min.	Typ.	Max.	Unit	
Conversion speed	F _c	V _{DD} = 4.75 to 5.25V T _a = -40 to +85°C V _{IN} = 0.5 to 2.5V f _{IN} = 1kHz ramp	0.5		20	MSPS	
Analog input band width (-1dB)	BW	Envelope		18		MHz	
Offset voltage*1	E _{OT}	Potential difference to V _{RT}	-60	-40	-20	mV	
	E _{OB}	Potential difference to V _{RB}	+20	+40	+60		
Integral non-linearity error	E _L	End point		+0.5	+1.3	LSB	
Differential non-linearity error	E _D			±0.3	±0.5		
Differential gain error	DG	NTSC 40 IRE mod ramp F _c = 14.3MSPS		1.0		%	
Differential phase error	DP			0.5		deg	
Aperture jitter	t _{aj}			30		ps	
Sampling delay	t _{sd}			4		ns	
Clamp offset voltage*2	E _{oc}	V _{IN} = DC, PWS = 3μs	V _{REF} = 0.5V	0	+20	+40	mV
			V _{REF} = 2.5V	-50	-30	-10	
Clamp pulse width (Sync pin input)	t _{cpw}	C = 100pF, R = 130kΩ (15PIN)	1.75	2.75	3.75	μs	
Clamp pulse delay	t _{cpd}			25		ns	

*1 The offset voltage E_{OB} is a potential difference between V_{RB} and a point of position where the voltage drops equivalent to 1/2 LSB of the voltage when the output data changes from "00000000" to "00000001".
E_{OT} is a potential difference between V_{RT} and a potential of point where the voltage rises equivalent to 1/2LSB of the voltage when the output data changes from "11111111" to "11111110".

*2 Clamp offset voltage varies individually. When using with R, G, B 3 channels, color sliding may be generated.

DC characteristics

($F_c = 20\text{MSPS}$, $V_{DD} = 5\text{V}$, $V_{RB} = 0.5\text{V}$, $V_{RT} = 2.5\text{V}$, $T_a = 25^\circ\text{C}$)

Item	Symbol	Conditions	Min.	Typ.	Max.	Unit
Supply current	I_{DD}	$F_c = 20\text{MSPS}$ NTSC ramp wave input		12	18	mA
Reference pin current	I_{REF}		4.5	6.6	8.7	mA
Analog input capacitance	C_{IN}	$V_{IN} = 1.5\text{V} + 0.07\text{V}_{rms}$		11		pF
Reference resistance (V_{RT} to V_{RB})	R_{REF}		230	300	450	Ω
Self-bias I	VRB_1	VRB and $VRBS$ are shorted V_{RT} and $VRTS$ are shorted	0.48	0.52	0.56	V
	VRT_1 to VRB_1		1.96	2.08	2.22	
Self-bias II	VRT_2	$VRB = AGND$ V_{RT} and $VRTS$ are shorted		2.32		V
Digital input voltage	V_{IH}	$V_{DD} = 4.75$ to 5.25V $T_a = -40$ to $+85^\circ\text{C}$	4.0			V
	V_{IL}				1.0	
Digital input current	I_{IH}	$V_{DD} = \text{max}$	$V_{IH} = V_{DD}$		5	μA
	I_{IL}		$V_{IL} = 0\text{V}$		5	
Digital output current	I_{OH}	$\overline{OE} = V_{SS}$	$V_{OH} = V_{DD} - 0.5\text{V}$	-1.1		mA
	I_{OL}	$V_{DD} = \text{min}$	$V_{OL} = 0.4\text{V}$	3.7		
	I_{OZH}	$\overline{OE} = V_{DD}$	$V_{OH} = V_{DD}$		16	μA
	I_{OZL}	$V_{DD} = \text{max}$	$V_{OL} = 0\text{V}$		16	

Timing

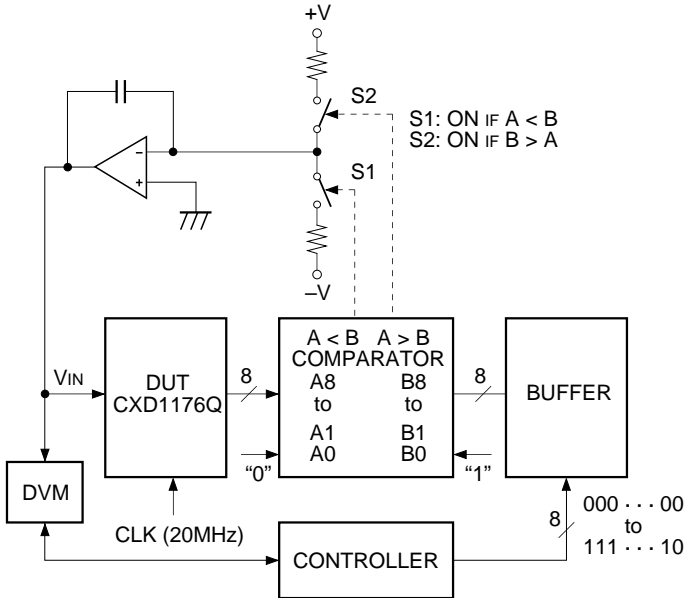
($F_c = 20\text{MSPS}$, $V_{DD} = 4.75$ to 5.25V , $V_{RB} = 0.5\text{V}$, $V_{RT} = 2.5\text{V}$, $T_a = -40$ to $+85^\circ\text{C}$)

Item	Symbol	Conditions	Min.	Typ.	Max.	Unit
Output data delay	T_{DL}	with TTL 1 gate and 10pF load		18	30	ns
Tri-state output enable time	t_{PZH}	$R_L = 1\text{k}\Omega$, $C_L = 20\text{pF}$ $\overline{OE} = 3\text{V} \rightarrow 0\text{V}$	2.5	6	10	ns
	t_{PZL}					
Tri-state output disable time	t_{PHZ}	$R_L = 1\text{k}\Omega$, $C_L = 20\text{pF}$ $\overline{OE} = 0\text{V} \rightarrow 3\text{V}$	8	18	30	ns
	t_{PLZ}					

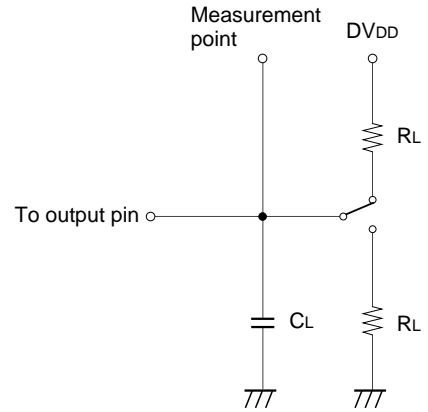
Electrical Characteristics Measurement Circuit

Integral non-linearity error
 Differential non-linearity error
 Offset voltage

} measurement circuit



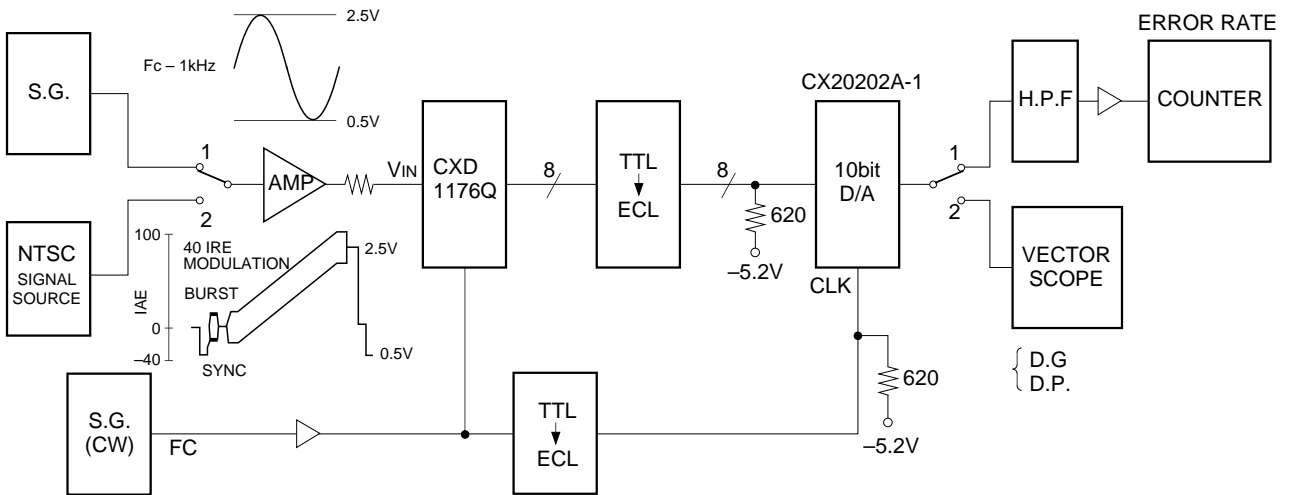
Tri-state output measurement circuit



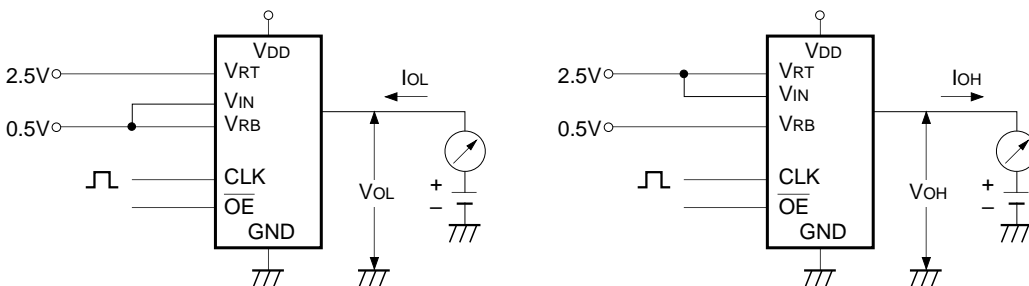
Note) CL includes capacitance of the probe and others.

Maximum operational speed
 Differential gain error
 Differential phase error

} measurement circuit

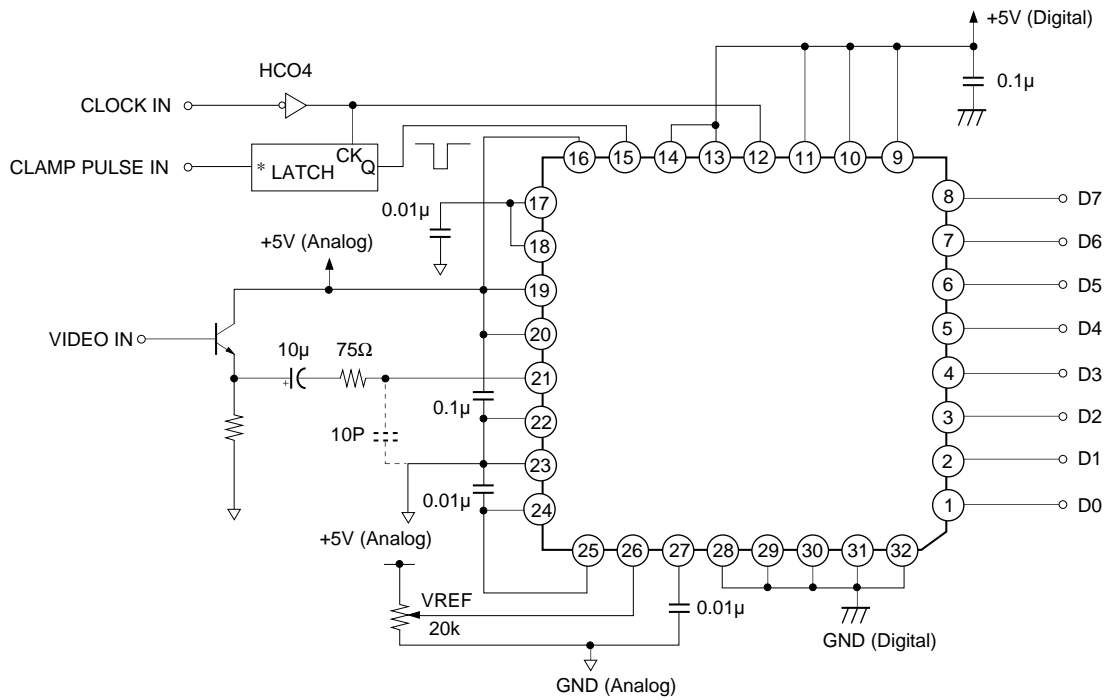


Digital output current measurement circuit

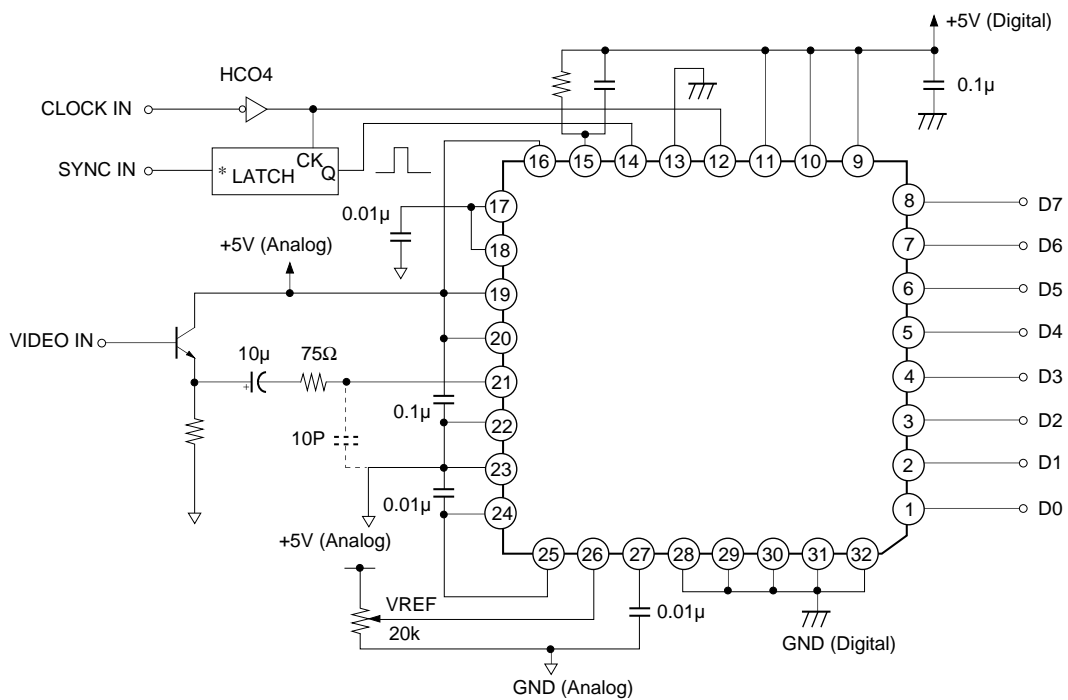


Application Circuit

(1) Case where clamp pulse is directry input (self-bias used)

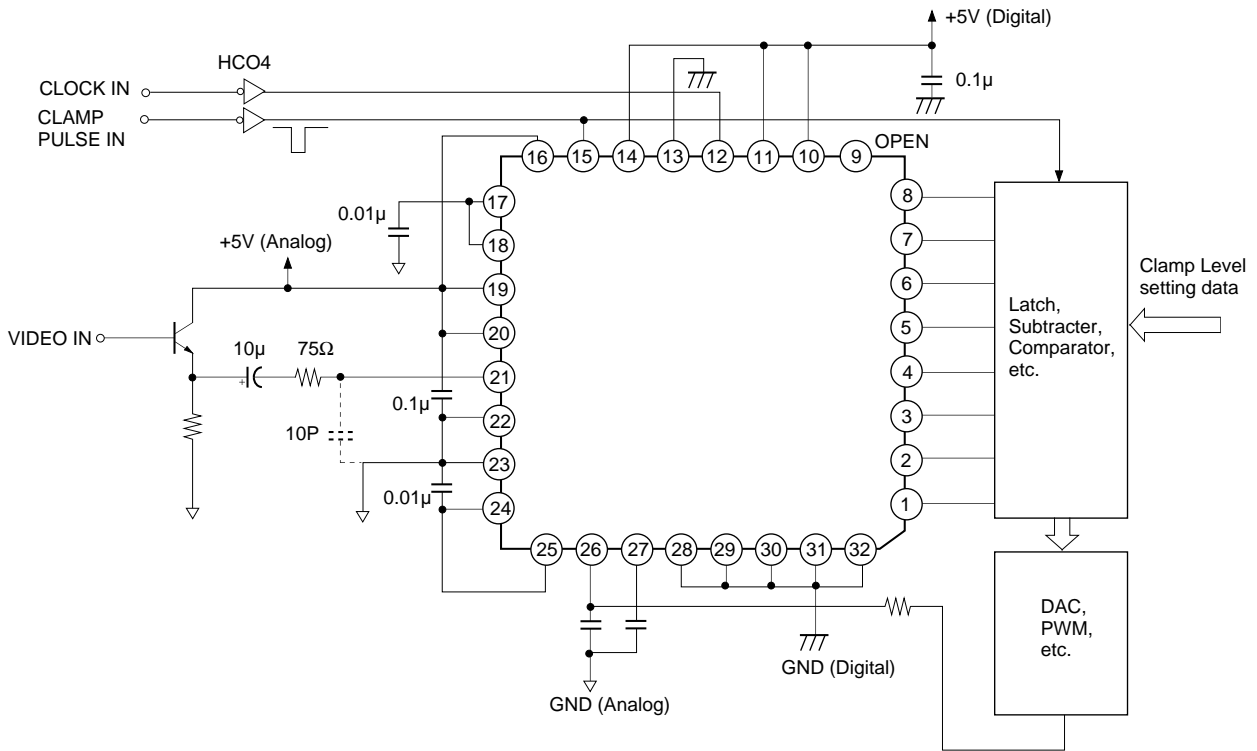


(2) Example where pedestal clamp is executed by sync pulse (self-bias used)

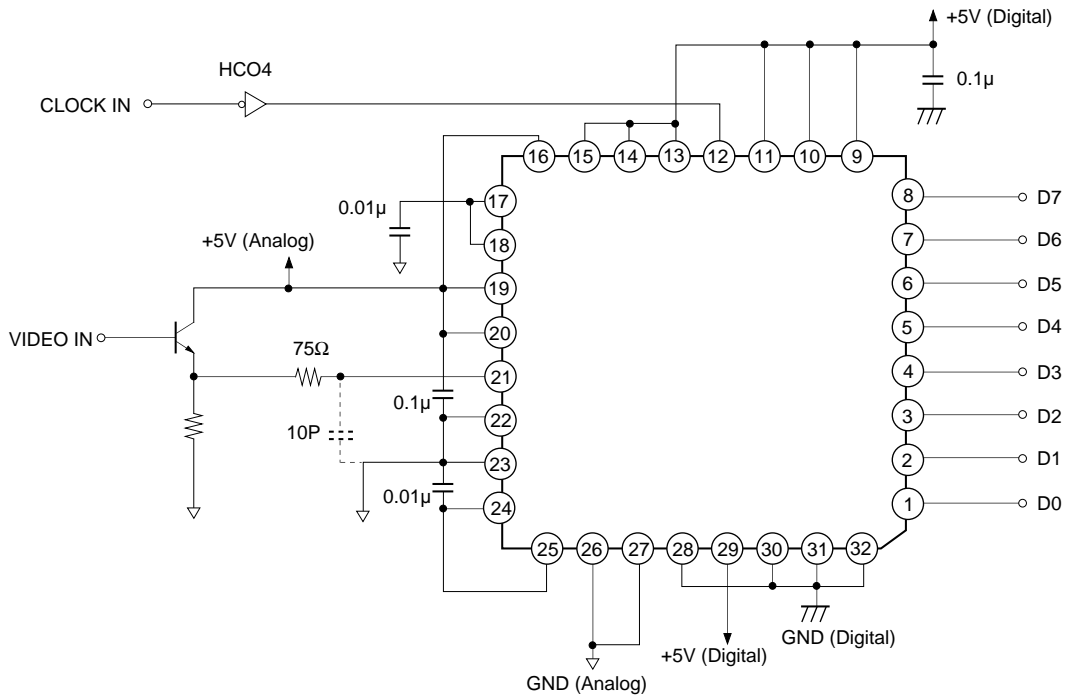


* The clamp pulse is latched by the ADC sampling clock, but that is not necessary for clamp basic operation. However, slight beat may be generated as vertical sag according to the relation between the sampling frequency and clamp pulse frequency. At such time, the latch circuit is effective. (See Notes on Operation (5).)

(3) Digital clamp (self-bias used)



(4) When clamp is not used (self-bias used)



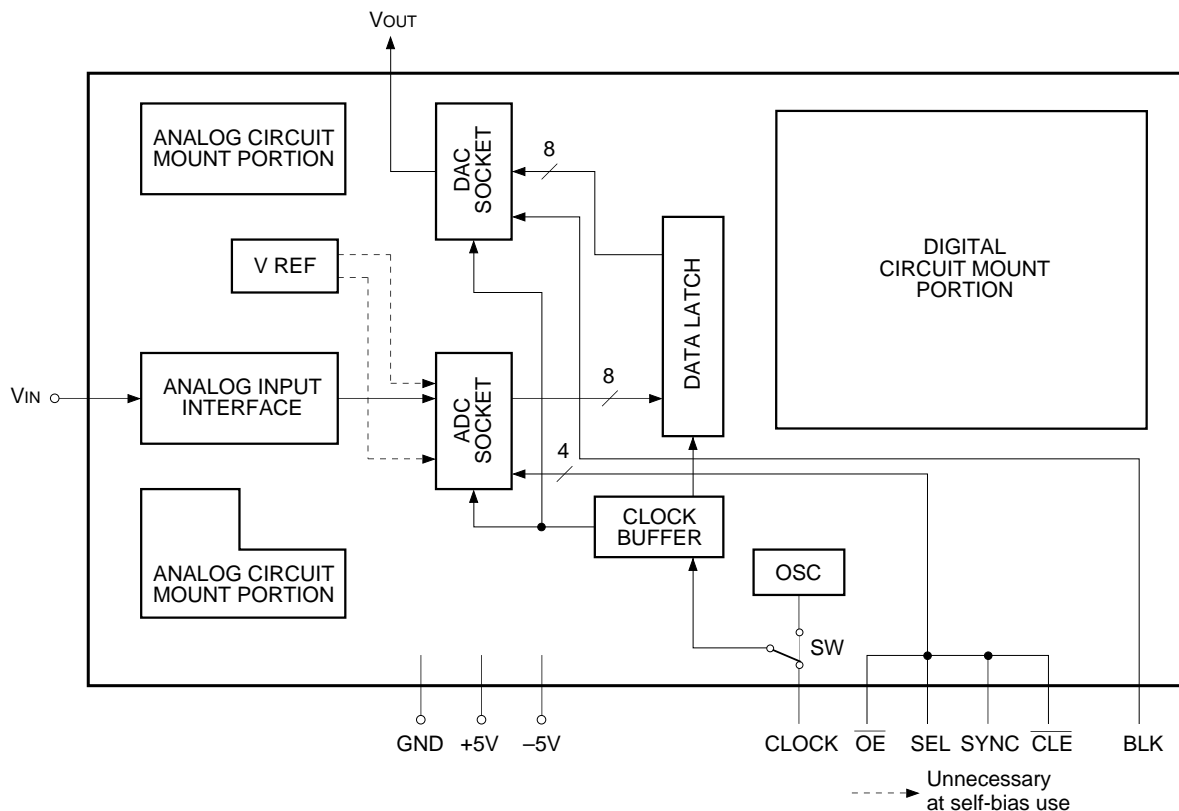
8-bit 20MSPS ADC and DAC Evaluation Board

Evaluation boards are available for the high speed, low power consumption CMOS converters, CXD1176Q (8-bit 20MHz A/D) and CXD1171M (8-bit 40MHz D/A).

The evaluation board is composed of a main board common to either type, to which is added sub board CXD1176Q or sub board CXD1171M. The junction is made through a socket.

To the main board are mounted an input interface, clock buffer and latch. To each of the sub boards is mounted CXD1176Q and CXD1171M respectively. Those IC's are mounted according to recommended print patterns designed to provide maximum performance to the A/D and D/A converters.

Block Diagram



Characteristics

- Resolution 8bit
- Maximum conversion rate 20MHz
- Digital input level CMOS level
- Supply voltage ±5.0V (Single +5V power supply possible at self bias use)

Supply voltage

Item	Min.	Typ.	Max.	Unit
+5V			150	mA
-5V			20	

Clock input

CMOS compatible

Pulse width	T _{cw1}	22.5ns (min)
	T _{cw0}	22.5ns (min)

Analog Output (CXD1171M) (RL > 10kΩ)

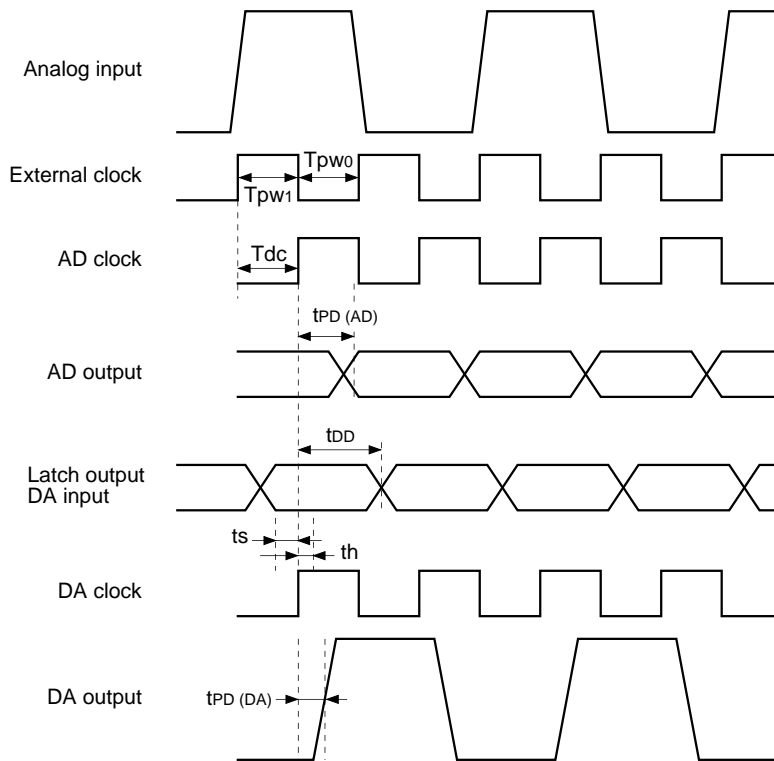
Item	Min.	Typ.	Max.	Unit
Analog output	1.9	2.0	2.1	V

Output Format (CXD1176Q)

The table shows the output format of AD Converter.

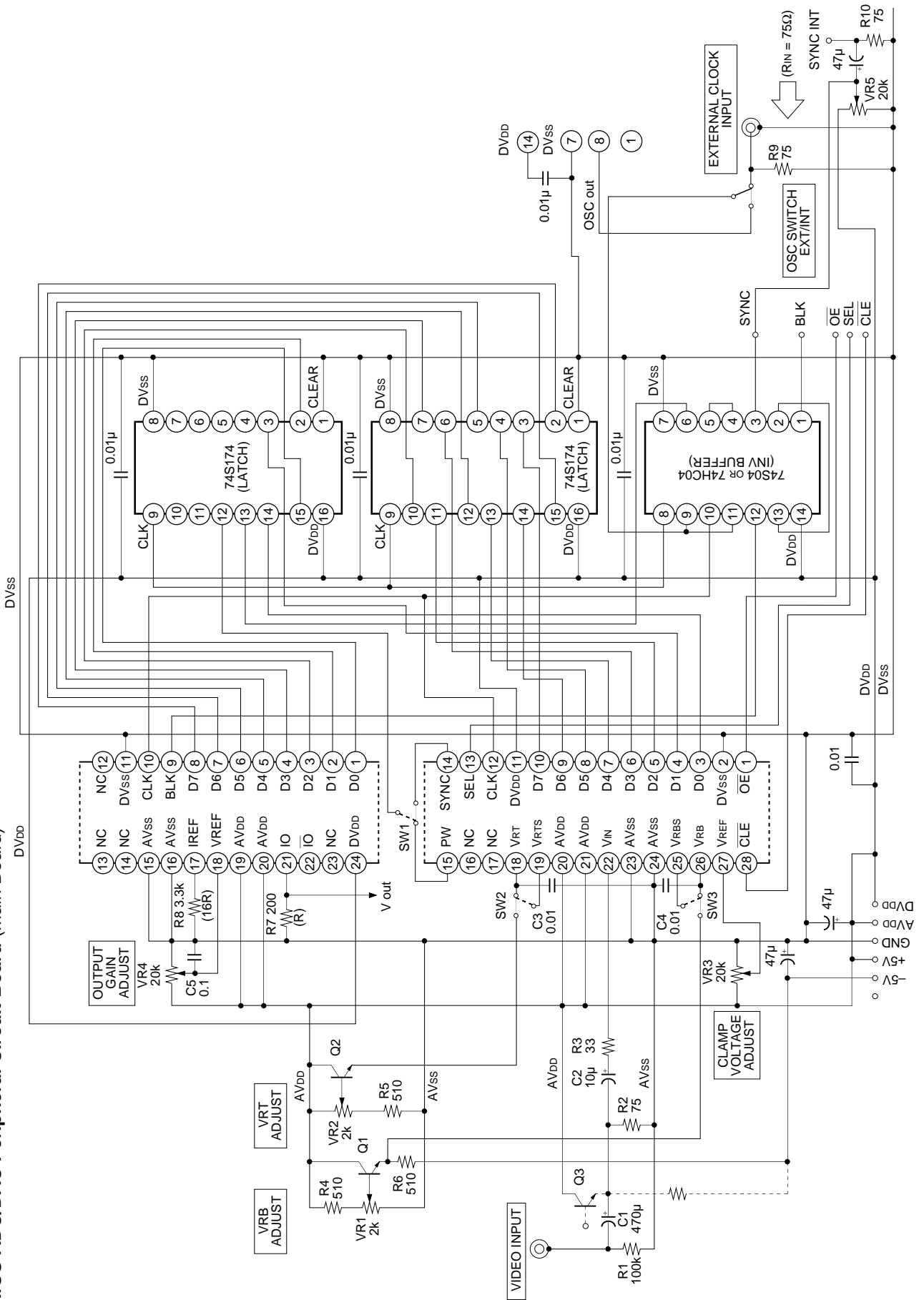
Analog input voltage	Step	Digital output code							
		MSB				LSB			
V _{RT}	0	1	1	1	1	1	1	1	1
⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮
⋮	127	1	0	0	0	0	0	0	0
⋮	128	0	1	1	1	1	1	1	1
⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮
V _{RB}	255	0	0	0	0	0	0	0	0

Timing Chart

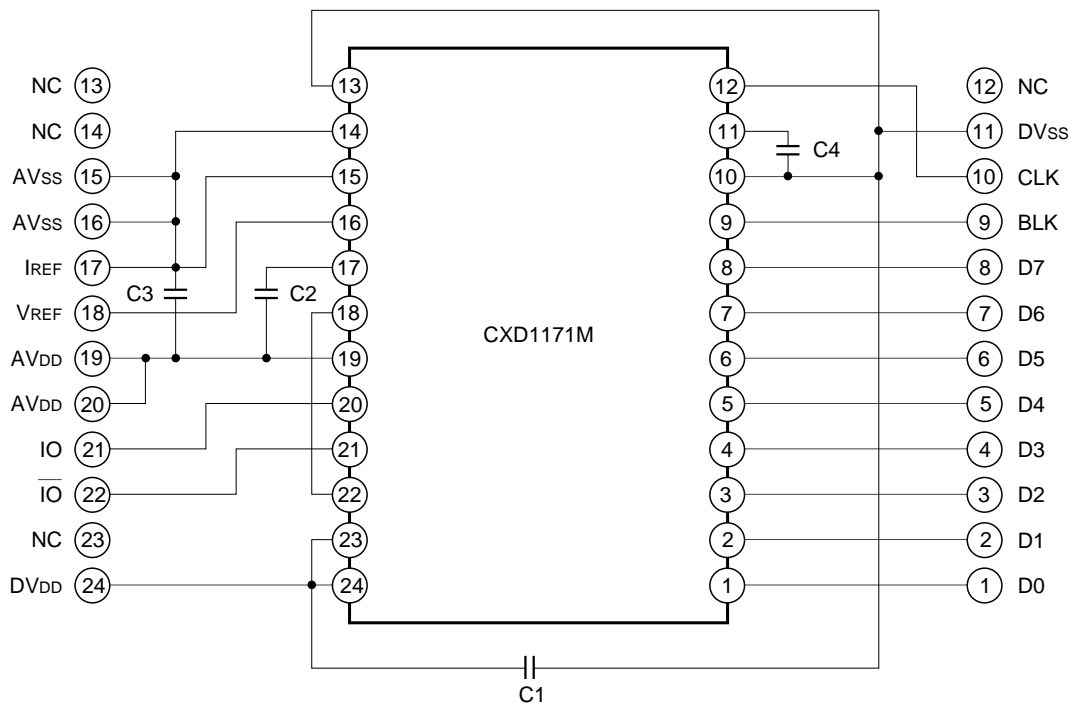
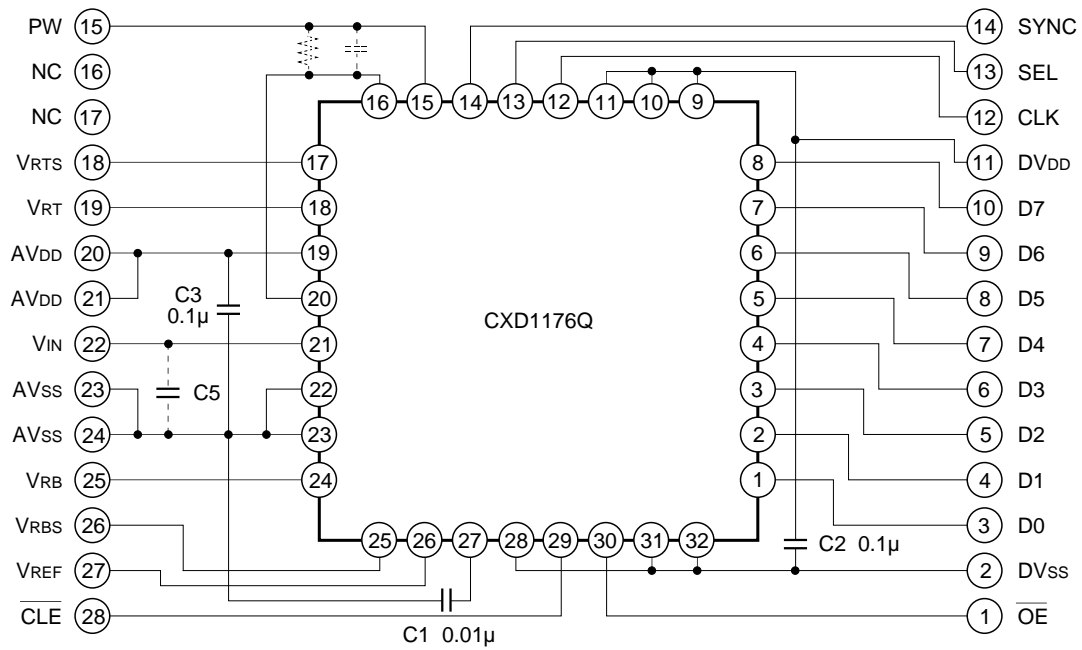


Item	Symbol	Min.	Typ.	Max.	Unit
Clock High time	T _{PW1}	25			ns
Clock Low time	T _{PW0}	25			ns
Clock Delay	T _{dc}			24	ns
Data delay AD	t _{PD (AD)}		18	30	ns
Data delay (latch)	t _{DD}			5	ns
Set up time	t _s	5			ns
Hold time	t _h	10			ns
Data delay DA	t _{PD (DA)}		10		ns

CMOS ADC/DAC Peripheral Circuit Board (Main Board)



CMOS ADC/DAC Peripheral Circuit Board (Sub Board)



List of Parts

resistance		transistor	
R1	100k Ω	Q1	2SC2785
R2	75 Ω	Q2	2SC2785
R3	75 Ω	Q3	2SC2785
R4	510 Ω		
R5	510 Ω	IC	
R6	510 Ω	IC1	74S174
R7	R = 200 Ω	IC2	74S174
R8	18R \approx 3.3k Ω	IC3	74S04
R9	75 Ω		
R10	75 Ω	oscillator	
VR1	2k Ω	OSC	
VR2	2k Ω		
VR3	20k Ω	others	
VR4	20k Ω	connector	BNC071
VR5	20k Ω	SW	AT1D2M3
capacitance			
C1	470 μ F/6.3V (chemical)		
C2	10 μ F/16V (chemical)		
C3	0.01 μ F		
C4	0.01 μ F		
C5	0.1 μ F		
C6	0.1 μ F		
C7	0.1 μ F		
C8	0.1 μ F		
C9	0.1 μ F		
C10	0.1 μ F		
C11	47 μ F/10V (chemical)		
C12	47 μ F/10V (chemical)		
C13	47 μ F/10V (chemical)		
C14	0.1 μ F		

Adjustment

1. Vref adjustment (VR1, VR2)
Adjustment of A/D converter reference voltage. VRB is adjusted through VR1 and VRT through VR2. When self-bias is used, there is no need for adjustment. Reference voltage is set through self-bias delivery.
2. Setting of clamp reference voltage (VR3)
Clamp reference voltage is set.
3. DAC output full-scale adjustment (VR4)
Full-scale voltage of D/A converter output is adjusted at the PCB shipment, the full-scale voltage is adjusted to approx. 2V.
4. Sync (clamp) pulse interface (VR5)
This adjustment enables interface with the signal generator and others at the PCB shipment, adjustment is performed to obtain a threshold of approx. 2.5V to an H sync of 0 to 5V.

5. \overline{OE} , SEL, Sync, BLK, \overline{CLE} , Sync INT

The following pins are set on the main board: \overline{OE} , SEL, Sync, \overline{CLE} , Sync INT (CXD1176Q) and BLK (CXD1171M). For the pins function, refer to the specifications. The difference between Sync pin and Sync INT pin is that you input a horizontal synchronizing signal above 3.5Vp-p Sync INT pin. The pulse threshold is set through VR5. For input through Sync pin, pulse is input at TTL or CMOS level. In this case cut off the junction line between Sync and Sync INT pin.

At the PCB shipment the main board pins are set as follows.

- \overline{OE} ... Low (A/D output ON)
- SEL ... Low (Pulse generated with Sync falling edge as trigger)
- Sync ... Line junction with Sync INT pin
- \overline{CLE} ... Low (Clamp function ON)
- BLK ... Low (Blanking OFF)

6. Clamp pulse input method

One method, as shown in Application Circuit examples (1) and (2), is to directly input the clamp pulse. The other is to use the built-in monostable multivibrator. The method is selected through SW1. At the PCB shipment it is set to direct input. To use the built-in monostable multivibrator, it is necessary to mount on the CXD1176Q sub board, R and C that determine pulse width.

(Ex. R = 130k, C = 100p, Tpw = 2.75 μ s Typ.)

Points on the PCB Pattern Layout

1. Set the layout not to have Digital current flow into Analog GND (Part 1). (For 1, See p. 21 Component side diagram.)
2. At CXD1176Q sub board, C₂ and C₃ capacitors serve the important role of bringing out CXD1176Q's full performance.
There are over 0.1 μ F (ceramic) capacitors with good high frequency characteristics. Layout as close to the IC as possible.
3. Analog GND (AV_{ss}) and Digital GND (DV_{ss}) are on a common voltage and power source. Keeping ADC's DV_{ss} (Part 2) as close as possible to the voltage supply source will provide better results. That is, a layout where ADC is close to the voltage supply source, is recommended. (For 2, see p. 21 Component side diagram.)
4. ADC samples analog signals at the clock falling edge point. Accordingly clocks supplied to ADC should not have any jitter.
5. The PCB layout shows ADC and DAC's Analog GND independently from the voltage generating source. On this PCB, the layout aims at providing an independent evaluation of ADC and DAC, as much as possible. On the actual board, common use will not cause any problems.

Notes on Operation

1. Reference voltage

Shorting V_{RT} and V_{RTS} , V_{RB} and V_{RBS} will activate the self-bias function that generates $V_{RT} = 2.6V$ and $V_{RB} = 0.5V$. On the PCB, either self-bias or the external reference voltage can be selected depending on the junction method of the jumper line. At shipment from the factory, reference voltage is provided in self-bias. Also, to provide external reference voltage, adjust the dynamic range ($V_{RT} - V_{RB}$) to above 1.8Vp-p.

2. Clock input

There are 2 modes for the PCB clock input

- 1) Provided from the external signal generator. (External clock)
- 2) Using the crystal oscillator (built-in clock driver). (Internal clock)

The 2 modes are selected using the switch on the PCB.

3. The 2 Latch IC's (74S174) are not absolutely necessary for the evaluation of ADC and DAC. That is, operation will still be normal if ADC output data is directly input to DAC input. However, as ADC output data is hardly ever D/A converted without executing Digital signal processing, it was mounted to indicate an example layout of Digital signal processing IC.

4. When clamp is not used

Turning CLE to H will set OFF the clamp function. In this case, the DC element is cut off by means of C_2 on the main board and DC voltage on the ADC side of C_2 turns to about $1/2 (V_{RT} + V_{RB})$. To transfer DC elements of input signals, short C_2 . At that time, it is necessary to bias input signals, but keeping R_2 open, Q_3 can also be used as buffer. Use the open space for the bias circuit.

5. Clamp pulse latch

On the evaluation board, the clamp pulse is latched with ADC sampling CLK and then input to either PW pin or Sync pin. This is to minimize V_{sag} due the synchronizing of noise and clamp pulse beat elements with GND sampling clock around ADC. If there are no problems with V_{sag} , latch is not necessary.

6. Peripheral through hole

There is a group of through holes on the Analog input, output and Logic. There are to be used when mounting additional circuits to the PCB. Use when necessary.

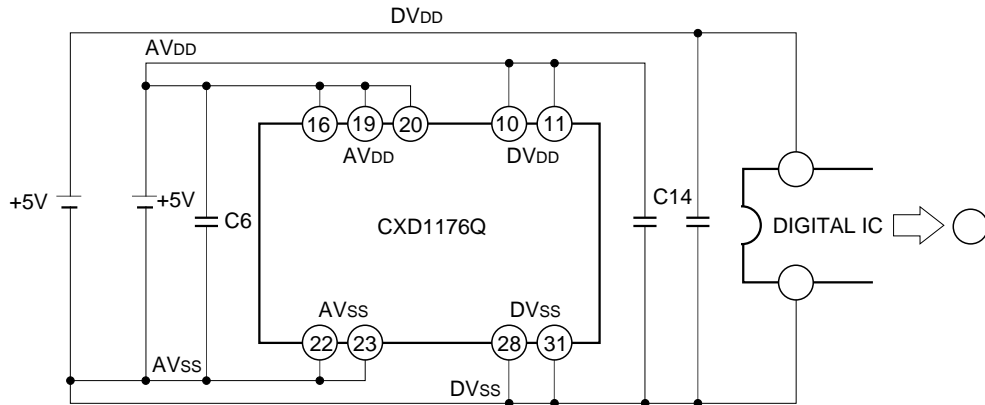
The connector hole on DAC part is used to mount the test chassis and the mount jack.

Latch Up Prevention

The CXD1176Q is a CMOS IC which requires latch up precautions. Latch up is mainly generated by the lag in the voltage rising time of AV_{DD} (Pins 16, 19 and 20) and DV_{DD} (Pin 10 and 11), when power supply is ON.

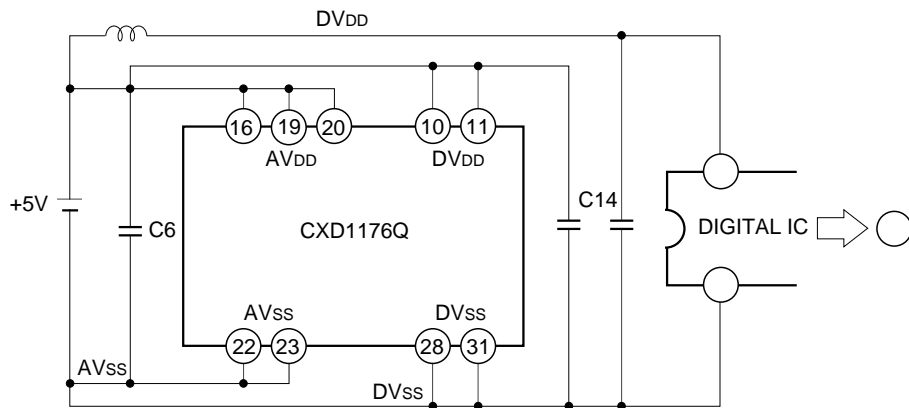
1. Correct usage

a. When analog and digital supplies are from different sources

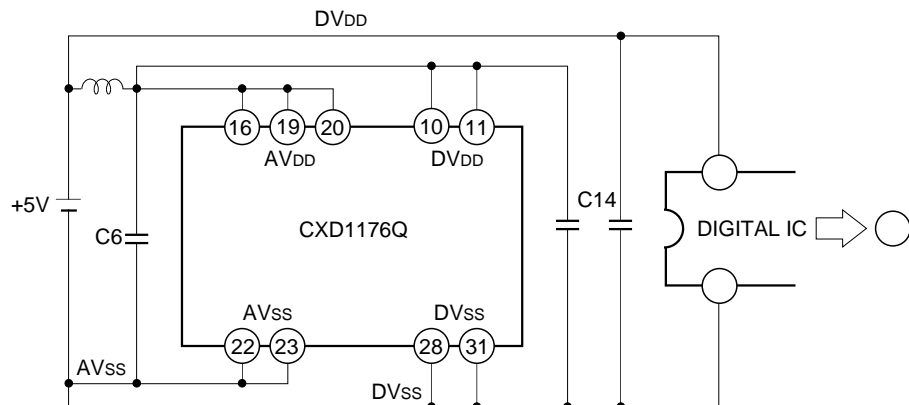


b. When analog and digital supplies are from a common source

(i)

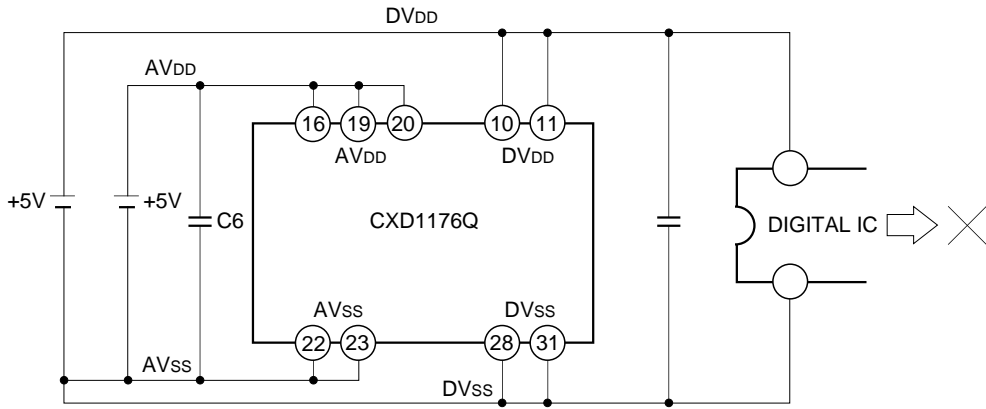


(ii)



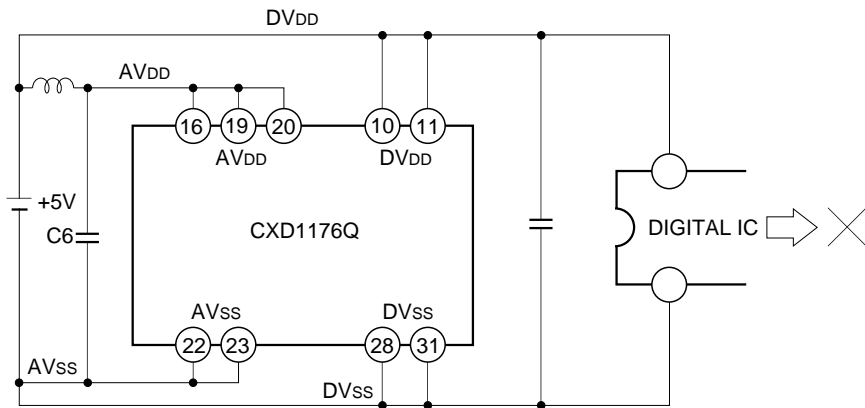
2. Example when latch up easily occurs

a. When analog and digital supplies are from different sources

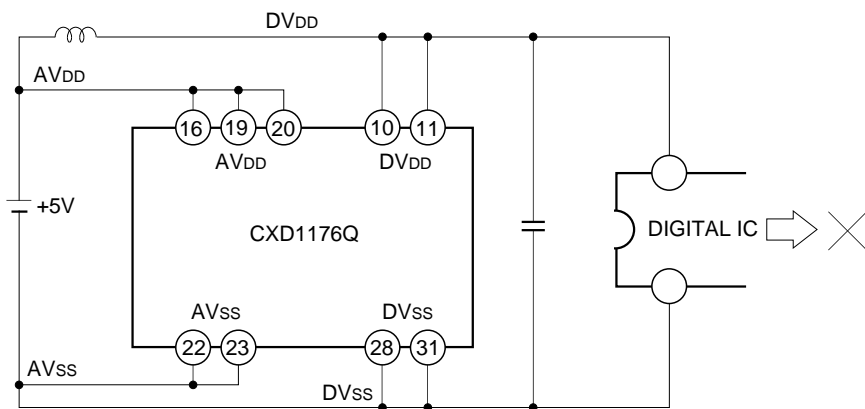


b. When analog and digital supplies are from a common source

(i)

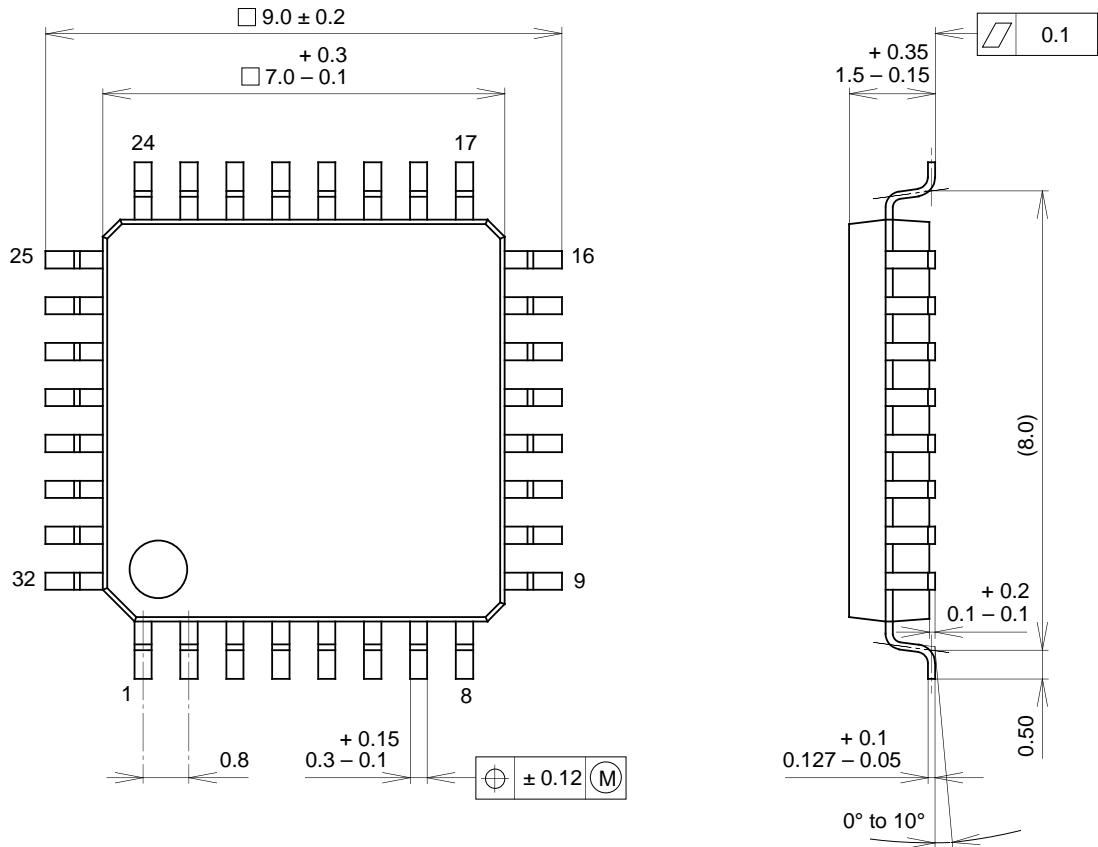


(ii)



Package Outline Unit: mm

32PIN QFP (PLASTIC)



SONY CODE	QFP-32P-L01
EIAJ CODE	*QFP032-P-0707-A
JEDEC CODE	_____

PACKAGE MATERIAL	EPOXY RESIN
LEAD TREATMENT	SOLDER PLATING
LEAD MATERIAL	42 ALLOY
PACKAGE WEIGHT	0.2g